

香港中文大學

The Chinese University of Hong Kong

THE CHINESE UNIVERSITY OF HONG KONG

DEPARTMENT OF MECHANICAL & AUTOMATION ENGINEERING

MAEG5070 Nonlinear Control Systems

Assignment #8

by

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Problem 1

Consider the controlled van del Pol equation

$$\begin{aligned}\dot{x}_1 &= x_2 \\ \dot{x}_2 &= -x_1 + \epsilon (1 - x_1^2) x_2 + u, \epsilon > 0 \\ y &= x_1\end{aligned}\tag{1}$$

- (a) Calculate the relative degree of the system.
- (b) Find a state feedback control law so that the equilibrium point at the origin of the closed-loop is globally asymptotically stable.

Solution:

(a)

$$\dot{y} = \dot{x}_1 = x_2 \tag{2}$$

$$\ddot{y} = \dot{x}_2 = -x_1 + \epsilon (1 - x_1^2) x_2 + u = \alpha(x) + \beta(x) u \tag{3}$$

Therefore, the relative degree of the system is 2.

- (b) The state feedback control law so that the equilibrium point at the origin of the closed-loop is globally asymptotically stable is shown as follows:

$$\begin{aligned}u &= \frac{y_d^{(n)} - \sum_{i=1}^n \alpha_i e^{(n-i)} - \alpha(x)}{\beta(x)} \\ &= -\alpha_1 \dot{e} - \alpha_2 e - \alpha(x) \\ &= -\alpha_1 \dot{y} - \alpha_2 y - \alpha(x) \\ &= -\alpha_1 x_2 - \alpha_2 x_1 + x_1 - \epsilon (1 - x_1^2) x_2\end{aligned}\tag{4}$$

Because $\rho = n$, the closed-loop system can always be made an asymptotically stable linear system.

Choosing $\alpha_1 = 1$ and $\alpha_2 = 2$, the control law becomes

$$u = -x_2 - x_1 \tag{5}$$

Problem 2

The motion equation of a single-link robot manipulator is given by

$$J\ddot{\theta} + MgL \sin \theta = u \quad (6)$$

- (a) Give the state space equation of (6) with $x_1 = \theta$, $x_2 = \dot{\theta}$, and $y = x_1$
- (b) Assume $J = 5$, $gL = 1$, and $M = 10$. Let $y_d(t)$ be a sufficiently smooth time function over $t \in [0, \infty)$. Let $e(t) = y(t) - y_d(t)$. Design a state feedback control law so that $e(t)$ satisfies $\ddot{e}(t) + 2\dot{e}(t) + e(t) = 0$.
- (c) Check your design in simple simulation for $y_d(t)$ to be a unit step input, and a sinusoidal function $\sin t$, respectively.

Solution:

- (a) The state space equation of (6) with $x_1 = \theta$, $x_2 = \dot{\theta}$, and $y = x_1$ is shown as follows

$$\begin{aligned} \dot{x}_1 &= x_2 \\ \dot{x}_2 &= -\frac{MgL \sin x_1}{J} + \frac{1}{J}u \\ y &= x_1 \end{aligned} \quad (7)$$

- (b) Because $J = 5$, $gL = 1$, and $M = 10$, the state space equation of (6) becomes

$$\begin{aligned} \dot{x}_1 &= x_2 \\ \dot{x}_2 &= -2 \sin x_1 + 0.2u \\ y &= x_1 \end{aligned} \quad (8)$$

Then, we will find the relative degree ρ :

$$\dot{y} = \dot{x}_1 = x_2 \quad (9)$$

$$\ddot{y} = \dot{x}_2 = -2 \sin x_1 + 0.2u = \alpha(x) + \beta(x)u \quad (10)$$

Therefore, the relative degree of the system is 2. state feedback control law so that the system is globally asymptotically stable

$$\begin{aligned} u &= \frac{y_d^{(n)} - \sum_{i=1}^n \alpha_i e^{(n-i)} - \alpha(x)}{\beta(x)} \\ &= \frac{\ddot{y}_d(t) - \alpha_1 \dot{e}(t) - \alpha_2 e(t) - \alpha(x)}{\beta(x)} \\ &= 5(\ddot{y}_d(t) - \alpha_1 \dot{e}(t) - \alpha_2 e(t) + 2 \sin x_1) \end{aligned} \quad (11)$$

Substituting Equation (11) into Equation (10) obtains

$$\ddot{y}(t) = \ddot{y}_d(t) - \alpha_1 \dot{e}(t) - \alpha_2 e(t) \quad (12)$$

That is

$$\ddot{e}(t) + \alpha_1 \dot{e}(t) + \alpha_2 e(t) = 0 \quad (13)$$

Therefore, $\alpha_1 = 2$ and $\alpha_2 = 1$ can satisfy the requirements of $\ddot{e}(t) + 2\dot{e}(t) + e(t) = 0$.

Hence, the state feedback control law is

$$u = 5(\ddot{y}_d(t) - 2(x_2 - \dot{y}_d) - (x_1 - y_d) + 2\sin x_1) \quad (14)$$

(c) We use the following Simulink to get the results:

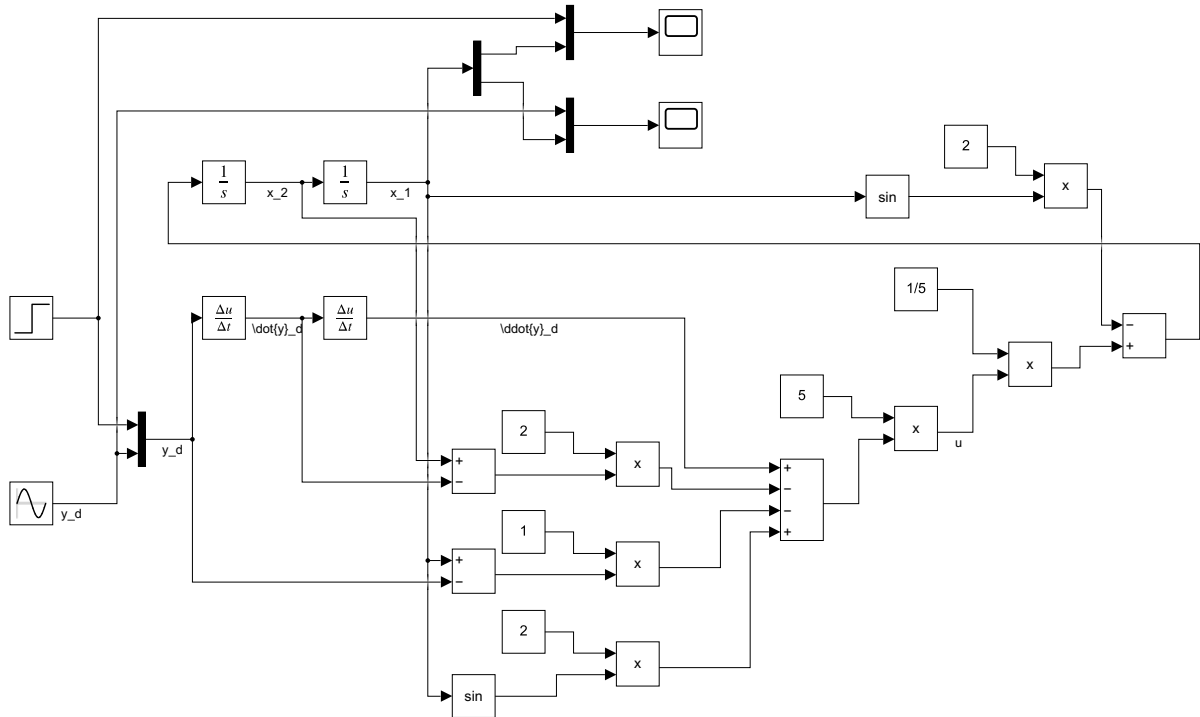


Figure 1: Block diagram for the system.

And we use the following code to plot the results:

```
1 clear all; clc;
2 figg1 = openfig('Q2Step.fig','reuse');
3 grid on;
4 xlabel('$t$, \mathrm{\left(s\right)}$', 'interpreter','latex');
5 ylabel('$y$, \mathrm{\left(m\right)}$', 'interpreter','latex');
6 legend('$y_d$', '$y$', 'interpreter','latex','Location','southeast');
7 title('');
8 a = get(gca,'XTickLabel');
9 set(gca,'XTickLabel',a,'FontName','Times','fontsize',12);
10 set(gcf,'renderer','painters');
11 filename = "Q2Step"+" .pdf";
12 saveas(gcf,filename);
13 close(figg1);
14 figg2 = openfig('Q2Sine.fig','reuse');
```

```

15 grid on;
16 xlabel('$t$, \mathrm{\left(s\right)}$', 'interpreter', 'latex');
17 ylabel('$y$, \mathrm{\left(m\right)}$', 'interpreter', 'latex');
18 legend('$y_d$', '$y$', 'interpreter', 'latex', 'Location', 'southeast');
19 title('');
20 a = get(gca, 'XTickLabel');
21 set(gca, 'XTickLabel', a, 'FontName', 'Times', 'fontsize', 12);
22 set(gcf, 'renderer', 'painters');
23 filename = "Q2Sine"+" .pdf";
24 saveas(gcf, filename);
25 close(figg2);

```

The results are shown as follows:

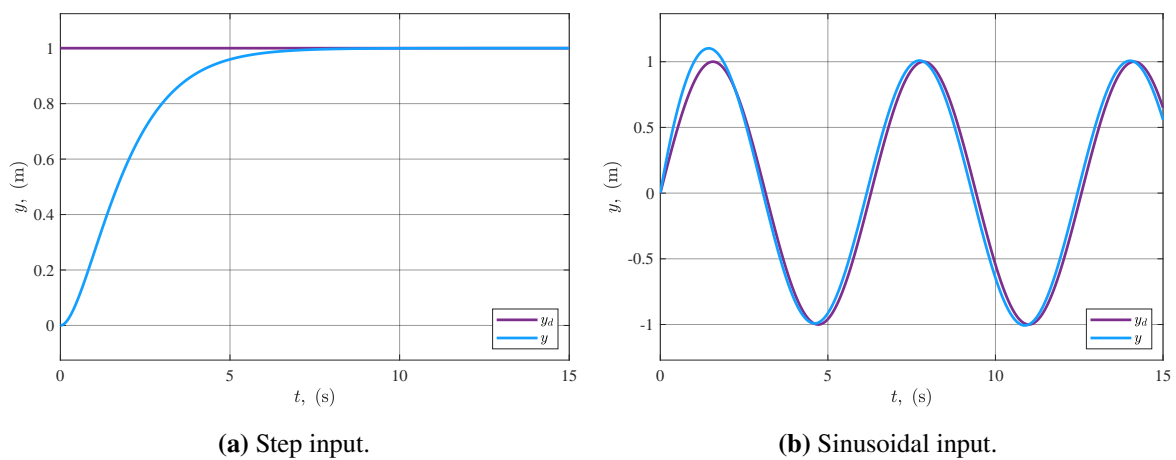


Figure 2: Simulation results.

Problem 3

Another way to achieve asymptotic tracking: Consider

$$y^{(n)} = \alpha(x) + \beta(x)u \quad (15)$$

or

$$\begin{aligned} \dot{x}_1 &= x_2 \\ &\dots \\ \dot{x}_{n-1} &= x_n \\ \dot{x}_n &= \alpha(x) + \beta(x)u \\ y &= x_1 \end{aligned} \quad (16)$$

where $x = [y \ \dot{y} \ \dots \ y^{(n-1)}]$, $\alpha(x)$ and $\beta(x)$ are known and $\beta(x) \neq 0$ for all x . Given $y_d(t)$, let $e(t) = y(t) - y_d(t)$ and define

$$s = e^{(n-1)} + \alpha_1 e^{(n-2)} + \dots + \alpha_{n-1} e \quad (17)$$

where $\alpha_1, \dots, \alpha_{n-1}$ are such that

$$\lambda^{n-1} + \alpha_1 \lambda^{n-2} + \dots + \alpha_{n-2} \lambda + \alpha_{n-1} \quad (18)$$

is a stable polynomial.

(a) Design a control law such that

$$\dot{s} + ks = 0 \quad (19)$$

where $k > 0$.

(b) Show that the control law achieves $\lim_{t \rightarrow \infty} e(t) = 0$.

(c) Show that, when $y_d = 0$, the closed-loop system is globally asymptotically stable.

Solution:

(a) To achieve asymptotic tracking for Equation (16), note that using an input transformation

$$\alpha(x) + \beta(x)u = u_a \quad (20)$$

or

$$u = \frac{u_a - \alpha(x)}{\beta(x)} \quad (21)$$

gives

$$y^{(n)} = u_a \quad (22)$$

which is in the chain integrator form. In order to achieve Equation (19) with $\rho = n$, i.e.,

$$\begin{aligned} & \left(y^{(n)} - y_d^{(n)} \right) + (\alpha_1 + k) \left(y^{(n-1)} - y_d^{(n-1)} \right) + (\alpha_2 + k\alpha_1) \left(y^{(n-2)} - y_d^{(n-2)} \right) \\ & + \dots + (\alpha_{n-1} + k\alpha_{n-2}) \left(y^{(1)} - y_d^{(1)} \right) + k\alpha_{n-1} (y - y_d) = 0 \end{aligned} \quad (23)$$

Substituting Equation (22) into (23) gives

$$\begin{aligned} & \left(u_a - y_d^{(n)} \right) + (\alpha_1 + k) \left(y^{(n-1)} - y_d^{(n-1)} \right) + (\alpha_2 + k\alpha_1) \left(y^{(n-2)} - y_d^{(n-2)} \right) \\ & + \dots + (\alpha_{n-1} + k\alpha_{n-2}) \left(y^{(1)} - y_d^{(1)} \right) + k\alpha_{n-1} (y - y_d) = 0 \\ \Rightarrow u_a &= y_d^{(n)} - \sum_{i=1}^n (\alpha_i + k\alpha_{i-1}) e^{n-i} \end{aligned} \quad (24)$$

Here $\alpha_0 = 1$ and $\alpha_n = 0$. Thus,

$$u = \frac{u_a - \alpha(x)}{\beta(x)} = \frac{y_d^{(n)} - \sum_{i=1}^n (\alpha_i + k\alpha_{i-1}) e^{n-i} - \alpha(x)}{\beta(x)} \quad (25)$$

(b) Because

$$s = e^{(n-1)} + \alpha_1 e^{(n-2)} + \dots + \alpha_{n-1} e \quad (26)$$

where $\alpha_1, \dots, \alpha_{n-1}$ are such that

$$\lambda^{n-1} + \alpha_1 \lambda^{n-2} + \dots + \alpha_{n-2} \lambda + \alpha_{n-1} \quad (27)$$

is a stable polynomial, for

$$\dot{s} = e^{(n)} + \alpha_1 e^{(n-1)} + \dots + \alpha_{n-1} e^{(1)} \quad (28)$$

where $\alpha_1, \dots, \alpha_{n-1}$ are such that

$$\lambda^n + \alpha_1 \lambda^{n-1} + \dots + \alpha_{n-2} \lambda^2 + \alpha_{n-1} \lambda \quad (29)$$

is also a stable polynomial by Routh-Hurwitz stability criterion. Therefore,

$$\begin{aligned} \dot{s} + ks &= e^{(n)} + (\alpha_1 + k) e^{(n-1)} + (\alpha_2 + k\alpha_1) e^{(n-2)} \\ &+ \dots + (\alpha_{n-1} + k\alpha_{n-2}) e^{(1)} + k\alpha_{n-1} e = 0 \end{aligned} \quad (30)$$

where $\alpha_i + k\alpha_{i-1}$, $i = 1, \dots, n$ are such that

$$\lambda^n + (\alpha_1 + k) \lambda^{n-1} + \dots + (\alpha_{n-1} + k\alpha_{n-2}) \lambda + k\alpha_{n-1} \quad (31)$$

is also a stable polynomial. As a result, e satisfies $\lim_{t \rightarrow \infty} e(t) = 0$.

(c) When $y_d = 0$, the closed-loop system is

$$\begin{aligned} \dot{x} &= Ax + Bk(x, 0, \dots, 0) = (A - B[k\alpha_{n-1}, (\alpha_{n-1} + k\alpha_{n-2}), \dots, (\alpha_1 + k)])x \\ &= \begin{bmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ \dots & & & & \\ 0 & 0 & 0 & \dots & 1 \\ -k\alpha_{n-1} & -(\alpha_{n-1} + k\alpha_{n-2}) & -(\alpha_{n-2} + k\alpha_{n-3}) & \dots & -(\alpha_1 + k) \end{bmatrix} x \end{aligned} \quad (32)$$

Clearly, $(A - B[k\alpha_{n-1}, (\alpha_{n-1} + k\alpha_{n-2}), \dots, (\alpha_1 + k)])$ is a companion matrix with its characteristic polynomial being

$$\lambda^n + (\alpha_1 + k)\lambda^{n-1} + \dots + (\alpha_{n-1} + k\alpha_{n-2})\lambda + k\alpha_{n-1} \quad (33)$$

Thus, when $y_d = 0$, the closed-loop system is globally asymptotically stable.